# Topic 3: Digital Sampling <br> Workshop Solutions 

## Workshop Questions

### 3.1 Two-dimensional Symmetry

Show that the DFT of a two-dimensional real function has the symmetry properties of

$$
\begin{aligned}
F_{R}(k, l) & =F_{R}(-k,-l) \\
F_{R}(-k, l) & =F_{R}(k,-l) \\
F_{I}(k, l) & =-F_{I}(-k,-l) \\
F_{I}(-k, l) & =-F_{I}(k,-l)
\end{aligned}
$$

## Solution

The two-dimensional DFT is given by

$$
F(k, l)=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} f(i, j) \exp \left(-l 2 \pi\left(\frac{k i}{N}+\frac{l j}{N}\right)\right)
$$

which can be written as:

$$
F(k, l)=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} f(i, j) \exp \left(-\imath 2 \pi \frac{k i}{N}\right) \exp \left(-\imath 2 \pi \frac{l j}{N}\right)
$$

which can then be expanded in terms of $\cos ()$ and $\sin ()$ to give

$$
F(k, l)=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} f(i, j)\left(\cos \left(2 \pi \frac{k i}{N}\right)-\imath \sin \left(2 \pi \frac{k i}{N}\right)\right)\left(\cos \left(2 \pi \frac{l j}{N}\right)-\imath \sin \left(2 \pi \frac{l j}{N}\right)\right)
$$

Now if $f(i, j)$ is real we can collect real and imaginary parts together to give

$$
F(k, l)=F_{R}(k, l)+\imath F_{I}(k, l)
$$

where

$$
F_{R}(k, l)=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} f(i, j)\left(\cos \left(2 \pi \frac{k i}{N}\right) \cos \left(2 \pi \frac{l j}{N}\right)-\sin \left(2 \pi \frac{k i}{N}\right) \sin \left(2 \pi \frac{l j}{N}\right)\right)
$$

and

$$
F_{I}(k, l)=\sum_{i=0}^{N-1} \sum_{j=0}^{N-1} f(i, j)\left(\cos \left(2 \pi \frac{k i}{N}\right) \sin \left(2 \pi \frac{l j}{N}\right)+\cos \left(2 \pi \frac{l j}{N}\right) \sin \left(2 \pi \frac{k i}{N}\right)\right)
$$

The symmetry relation is given by the $\cos ()$ and $\sin ()$ functions. Substitute,

$$
a=2 \pi \frac{k i}{N} \quad \& \quad b=2 \pi \frac{l j}{N}
$$

so the symmetry for the real part is given by:

$$
\cos (a) \cos (b)-\sin (a) \sin (b)
$$

so that, noting that $\cos (a)=\cos (-a)$ and $\sin (a)=-\sin (-a)$,

$$
F_{R}(a, b)=F_{R}(-a,-b) \quad \& \quad F_{R}(-a, b)=F_{R}(a,-b)
$$

and the imaginary part by:

$$
\cos (a) \sin (b)+\sin (a) \cos (b)
$$

so that:

$$
F_{I}(a, b)=-F_{I}(-a,-b) \quad \& \quad F_{I}(-a, b)=-F_{I}(a,-b)
$$

as expected.

### 3.2 Symmetry Pairing

Verify for a $6 \times 6$ image that the DFT of a two-dimensional real function has:

$$
\begin{array}{ll}
\frac{N^{2}}{2}+2 & \text { Independent real values } \\
\frac{N^{2}}{2}-2 & \text { Independent imaginary values }
\end{array}
$$

Fourier filters involve multiplying the DFT by a filtering function $H(i, j)$. Many of these filters are real only. Suggest a scheme for packing the real and imaginary parts of a DFT into a square array that makes multiplication with such a filter simple.

## Solution

First renumber the Fourier array so that the $(k: l=-3, \rightarrow, 2)$ which can be done due to the cyclic properties of the DFT. There are now 2 arrays of 36 element, one for the real parts and one form the imaginary parts as shown below.


Real Part: is symmetric. Consider three regions of the output place, shaded grey, black \& white above.

1. grey: There are 24 points where the symmetric symmetry is simple: eg: $F_{R}(1,2)=$ $F_{R}(-1,-2)$. So in this regions there are $\mathbf{1 2}$ unique values with the other 12 given by symmetry.
2. black: There are 4 points with no symmetric pair, actually due to the cylcic properties they are their own symmetric pair. This gives $\mathbf{4}$ uniques values.
3. white: There are 8 points in 4 symmetric pairs but they are wrapped round due to cyclic property. For example $F_{R}(-3,2)=F_{R}(3,-2)$ but $F_{R}(3,-2)$ is cyclic with period 6 , so that $F_{R}(3,-2)=F_{R}(-3,-2)$. This region gives 4 uniques values with the other 4 given by symmetry.

This gives total of:

| Area | Number |
| :--- | ---: |
| grey | 12 |
| black | 4 |
| white | 4 |
| Total | $\mathbf{2 0}$ |

Imaginary Part: is anit-symmetric. Consider same three regions,

1. grey: As for real there are 24 points where symmetry is simple: eg: $F_{I}(1,2)=-F_{I}(-1,-2)$.

So in this regions there are $\mathbf{1 2}$ unique values with the other 12 given by symmetry.
2. black: There are 4 points where all terms in the expression for the imaginary componets contain $\sin ( \pm n \pi)=0$. So all these 4 points are always zero and to not depend on $f(i, j)$ This area contributes $\mathbf{0}$ values.
3. white: There are 8 points in 4 symmetric pairs but they are wrapped round due to cyclic property. This region gives 4 uniques values with the other 4 given by symmetry.

This gives total of:

| Area | Number |
| :--- | ---: |
| grey | 12 |
| black | 0 |
| white | 4 |
| Total | $\mathbf{1 6}$ |

So as noted we have

$$
\begin{aligned}
& \frac{N^{2}}{2}+2=20 \quad \text { Real Values } \\
& \frac{N^{2}}{2}-2=16 \quad \text { Imaginary Values }
\end{aligned}
$$

There are 36 values that can be "packed" into a single $6 \times 6$ float array. This was shown for $N=4$ in lectures, $N=6$ here, and can (easily) be extended to any even $N$.
If we want to apply a real filter $H(k, l)$ then we need to multiply both the real and imaginary parts by the same filter to give:

$$
G_{R}(k, l)=F_{R}(k, l) H(k, l) \quad \& \quad G_{I}(k, l)=F_{I}(k, l) H(k, l)
$$

we can then inverse transform to give the filtered image. (For details of filter types see lecture 8). In order to preserve the symmetry of $G(k, l)$ the filter $H(k, l)$ must be symmetric. If we then pack the Fourier image data $F(k, l)$ as shown on left, with the

1. Real Symmetric part in the Dark Grey region.
2. Imaginary Anti-symmetric part in the Light Grey region.
3. Additional 4 Real Values in the Black regions.


Then if we apply a symmetric filter, as shown on the right, then the apperant multiplication reduced to a $N \times N$ real multiplication with the "top" half of the filter multiplying the real section and the "bottom" half the imaginary section.

### 3.3 Shifting The Centre

Show that if your two dimensional DFT code locates the $(0,0)$ term in the top/left of the array, then this can be shifted to the centre of the array by pre-multiplying the by a $\pm 1$ checker-board.

## Solution

Want to shift the top-left pixel to the centre of $N \times N$ array which occurs at ( $N / 2, N / 2$ ). Mathematically a shift can be implemented as a convolution with a shifted $\delta$-function, so we want to form:

$$
F(k, l) \odot \delta\left(k-\frac{N}{2}, l-\frac{N}{2}\right)
$$

(Note that we want cyclic wrap-round so that sections of $F(i, j)$ with $i, j>N / 2$ will wrap-round to the $i, j<N / 2$ sections. This will occur automatically due to the cyclic property of the DFT.) The convolution in Fourier Space is equivalent to a multiplication in Real space of

$$
f(i, j) \mathcal{F}\left\{\delta\left(k-\frac{N}{2}, l-\frac{N}{2}\right)\right\}
$$

We have to take the DFT, so

$$
\mathcal{F}\{\delta(k-a)\}=\frac{1}{N} \sum_{i=0}^{N-1} \delta(k-a) \exp \left(-\imath 2 \pi \frac{i k}{N}\right)=\frac{1}{N} \exp \left(-\imath 2 \pi \frac{a i}{N}\right)
$$

so in two dimensions we have that:

$$
\mathcal{F}\left\{\delta\left(k-\frac{N}{2}, l-\frac{N}{2}\right)\right\}=\frac{1}{N^{2}} \exp (\imath \pi(i+j))
$$

Ignoring the $1 / N^{2}$ term, we note that

$$
\begin{aligned}
\exp (\imath \pi(i+j)) & =1 \quad \text { when }(i+j) \text { Even } \\
& =-1 \quad \text { when }(i+j) \text { Odd }
\end{aligned}
$$

so for $N$ even we have to multiply the function $f(i, j)$ by

| 1 | -1 | 1 | $\cdots$ | -1 |
| :---: | :---: | :---: | :---: | :---: |
| -1 | 1 | -1 | $\cdots$ | 1 |
| $\vdots$ | $\vdots$ | $\vdots$ | $\ddots$ | $\vdots$ |
| 1 | -1 | 1 | $\cdots$ | -1 |
| -1 | 1 | -1 | $\cdots$ | 1 |

### 3.4 Speed of the FFT

On a particular computer system the FFT of a $128 \times 128$ image takes 0.11 seconds, estimate how long this system would take to calculate the FFT of a $1024 \times 1024$ image.

## Solution

For a one-dimensional FFT the computational cost is proportional to $N \log _{2}(N)$. The twodimensional DFT can be formulated at $2 N$ one-dimensional FFT ( $N$ along the rows, plus $N$ down the columns). The computational cost for the FFT of an $N \times N$ array is therefore

$$
\propto N^{2} \log _{2}(N)
$$

If it takes 0.11 secs for $N=128$ then for $N=1024$ is will take

$$
\frac{1024^{2} \log _{2}(1024)}{128^{2} \log _{2}(128)} \times 0.11=10.05 \mathrm{secs}
$$

Real experimental values for $N=2^{n}$ are:

| Size | Time in secs. |
| :--- | ---: |
| 64 | 0.02 |
| 128 | 0.11 |
| 256 | 0.46 |
| 512 | 2.31 |
| 1024 | 9.55 |

which is very close to the predicted timings. Experimental timings for various $N$ consisting of powers or $2 \& 3$ from $64 \rightarrow 1024$ are shown in the graph below.


The best fit of $N^{2} \log _{2}(N)$ is also plotted. The $N=2^{n}$ points are all below the curve while the values of $N$ containing 3(s) are above the line. This shows that this algorithm is optimally efficient for powers-of-two.
Timings for larger $N$ showed very significant deviation for expected $N^{2} \log _{2}(N)$ relation due to size of physical memory, for example a $2048 \times 2048$ DFT took 20 minutes.

### 3.5 CCD Sensors

A CCD sensor is a two-dimensional array of detectors that can be used to sample an image. A typical TV quality CCD camera will have $586 \times 768$ sensors on a 15 by 20 mm area with a $3: 4$. Calculate the size of the sensors and the maximum spatial frequency in the detected image.
You wish to use this CCD camera to image pages of text for a character recognition system that is able to easily resolved 8 pt ( 1 pt is $1 / 72 \mathrm{nd}$ of an inch) letters. What magnification is required and how large a page of text can be images at once.
Hint: To easily resolve a letter you must be able to resolve line approximately 5 times closer together than the minimum separation of lines in the letter.

## Solution

Video images have a 4:3 aspect ratio due to the shape of the normal TV screen. The sensor sizes are simply

$$
\frac{15}{586}=25.6 \mu \mathrm{~m} \quad \frac{20}{768}=26.0 \mu \mathrm{~m}
$$

so, for all practical purposes, they are square.
Shannon sampling rate is that the signal is optimally sampled with sampling interval $\Delta x$ if

$$
\Delta x=\frac{1}{W}
$$

where $W$ is the "bandwidth" of the signal.


The maximum spatial frequency is thus half the bandwidth, so that

$$
w_{0}=\frac{1}{2} W=\frac{1}{2 \Delta x}
$$

so the maximum spatial frequencies are, $u_{0}, v_{0}$,

$$
u_{0}=19.2 \text { lines } / \mathrm{mm} \quad \& \quad v_{0}=19.5 \text { lines } / \mathrm{mm}
$$

Letter in an 8 pt font are approximately 2.8 mm high. This is typically the smallest font that appears in printed documents.


The typical lines in the letter are about $1 / 5$ th of the letter's height, but to "easily" resolve the letter we must be able to resolve lines about 5 times closer together than that so we must resolve lines with a separation of

$$
\frac{2.8}{25}=112 \mu \mathrm{~m}
$$

which is equivalent to spatial frequency of:

$$
\text { 8.93lines } / \mathrm{mm}
$$

The CCD sensor has a maximum spatial frequency resolution of approximately 19 lines $/ \mathrm{mm}$, so the magnification between object and image plane is given by

$$
\frac{8.93}{19.2}=0.46
$$

in other words the imaged area of the page is approximately $\times 2.15$ the size of the sensor, so the size of the imaged "page" is approximately,

$$
43 \times 32 \mathrm{~mm}
$$

which is only a small section of a page. In a typical book an area of this size contains about 45 words.
This calculation shows that optical character recognition requires a vast amount of data to be processed. Real character recognition systems do not use video cameras by a linear sensor array the width of the paper.

